





On solvability of CGCARE for LQR problems with zero input-cost Chayan Bhawal, Debasattam Pal December 13, 2019 CSC-MPI Magdeburg and IIT Bombay



Consider a controllable system with state-space dynamics

$$\frac{d}{dt}x(t) = Ax(t) + Bu(t), \text{ where } A \in \mathbb{R}^{n \times n}, B \in \mathbb{R}^{n \times m}.$$



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Infinite-horizon linear quadratic regulator (LQR) problem

For every initial condition $x_0 \in \mathbb{R}^n$, find an input u(t) (from admissible input space) that minimizes the functional

$$J(x_0, u(t)) := \int_0^\infty \begin{bmatrix} x(t) \\ u(t) \end{bmatrix}^T \begin{bmatrix} Q & S \\ S^T & R \end{bmatrix} \begin{bmatrix} x(t) \\ u(t) \end{bmatrix} dt, \text{ where } \begin{bmatrix} Q & S \\ S^T & R \end{bmatrix} \ge 0$$

and $\lim_{t\to\infty} x(t) = 0$.



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and $\lim_{t\to\infty} x(t) = 0$.

For R > 0 (Regular case)

 $A^{T}K + KA + Q - (KB + S)R^{-1}(B^{T}K + S^{T}) = 0 \qquad u(t) = -R^{-1}(B^{T}K_{\max} + S^{T})x(t)$



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and $\lim_{t\to\infty} x(t) = 0.$

For $R \ge 0$, $\det(R) = 0$ (Singular/degenerate case)

 $A^{T}K + KA + Q - (KB + S)R^{-1}(B^{T}K + S^{T}) = 0 \qquad u(t) = -R^{-1}(B^{T}K_{\max} + S^{T})x(t)$



COMPUTATIONAL METHODS IN SYSTEMS AND CONTROL THEORY Literature review

System:
$$\frac{d}{dt}x = Ax + Bu$$
 Cost matrix: $\begin{bmatrix} Q & S \\ S^T & R \end{bmatrix} \ge 0, \det(R) = 0.$



COMPUTATIONAL METHODS IN SYSTEMS AND CONTROL THEORY LITERATURE REVIEW

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Theorem (Ferrante and Ntogramatzidis, Automatica 2014)
Singular LQR problem is solvable using a static state-feedback controller
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there exists $K = K^T \in \mathbb{R}^{n \times n}$ such that
 $A^T K + KA + Q - (KB + S)R^{\dagger}(B^T K + S^T) = 0,$
 $\ker(R) \subseteq \ker(S + KB).$



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Constrained Generalized Continuous Algebraic Riccati Equation - CGCARE.



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Singular LQR problems corresponding to single-input sytems can be solved using proportional-derivative (PD) controllers.



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When is CGCARE solvable?



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When is CGCARE solvable for R = 0?



Hamiltonian pencils COMPUTATIONAL METHODS IN SYSTEMS AND CONTROL THEORY

• For
$$R = 0$$
, $\begin{bmatrix} Q & S \\ S^T & R \end{bmatrix} \ge 0 \Rightarrow Q \ge 0$ and $S = 0$.

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Hamiltonian pencils

• Cost function: $\int_0^\infty (x^T Q x) dt$.



COMPUTATIONAL METHODS IN Hamiltonian pencils

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Pontryagin's Maximum principle: (x: states, z: costates, u: input)



COMPUTATIONAL METHODS IN SYSTEMS AND CONTROL THEORY HAMILTONIAN PENCILS

• Cost function: $\int_0^\infty (x^T Q x) dt$.

Pontryagin's Maximum principle: (x: states, z: costates, u: input)

$$\underbrace{\begin{bmatrix} I_{n} & 0 & 0\\ 0 & I_{n} & 0\\ 0 & 0 & 0 \end{bmatrix}}_{E} \frac{d}{dt} \begin{bmatrix} x\\ z\\ u \end{bmatrix} = \underbrace{\begin{bmatrix} A & 0 & B\\ -Q & -A^{T} & 0\\ 0 & B^{T} & 0 \end{bmatrix}}_{H} \begin{bmatrix} x\\ z\\ u \end{bmatrix}$$



COMPUTATIONAL METHODS IN Hamiltonian pencils

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(E, H): Hamiltonian matrix pair, (sE - H): Hamiltonian matrix pencil.



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Output-nulling representation:

$$\frac{d}{dt} \begin{bmatrix} x \\ z \end{bmatrix} = \underbrace{\begin{bmatrix} A & 0 \\ -Q & -A^T \end{bmatrix}}_{\widehat{A}} \begin{bmatrix} x \\ z \end{bmatrix} + \underbrace{\begin{bmatrix} B \\ 0 \end{bmatrix}}_{\widehat{B}} u, \text{ and } 0 = \underbrace{\begin{bmatrix} 0 & B^T \end{bmatrix}}_{\widehat{C}} \begin{bmatrix} x \\ z \end{bmatrix}$$



COMPUTATIONAL METHODS IN CGCARE solvability

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COMPUTATIONAL METHODS IN CGCARE solvability

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Lemma

CGCARE with
$$R = 0$$
 solvable
 $\widehat{}$
 $\mathcal{P}(s) := \widehat{C}(sI_{2n} - \widehat{A})^{-1}\widehat{B} \equiv 0$ (as a rational matrix).

For the R singular case: more necessary and sufficient conditions in Bhawal, Qais, and Pal, IEEE L-CSS 2019.



Autonomy?

$$\widehat{C}(sI_{2n} - \widehat{A})^{-1}\widehat{B} \equiv 0 \Leftrightarrow \widehat{C}\widehat{A}^k\widehat{B} = 0 \text{ for all } k \in \{0, 1, 2, \cdots\}.$$



CONTROL THEORY Autonomy?

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$$\widehat{C}(sI_{2n} - \widehat{A})^{-1}\widehat{B} \equiv 0 \Leftrightarrow \widehat{C}\widehat{A}^k\widehat{B} = 0$$
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• Note that

$$\widehat{C}\widehat{B} = \begin{bmatrix} 0 & B^T \end{bmatrix} \begin{bmatrix} B \\ 0 \end{bmatrix} = 0.$$



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COMPUTATIONAL METHODS IN SYSTEMS AND CONTROL THEORY

$$\begin{split} \widehat{C}\widehat{B} &= \begin{bmatrix} 0 & B^T \end{bmatrix} \begin{bmatrix} B \\ 0 \end{bmatrix} = 0. \\ \widehat{C}\widehat{A}\widehat{B} &= \begin{bmatrix} 0 & B^T \end{bmatrix} \begin{bmatrix} A & 0 \\ -Q & -A^T \end{bmatrix} \begin{bmatrix} B \\ 0 \end{bmatrix} = -B^TQB = 0 \Rightarrow QB = 0. \end{split}$$

Autonomy?



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$$\widehat{C}(sI_{2n} - \widehat{A})^{-1}\widehat{B} \equiv 0 \Leftrightarrow \widehat{C}\widehat{A}^k\widehat{B} = 0$$
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 $\widehat{C}\widehat{A}^k\widehat{B}=0 \Rightarrow (A^kB)^TQA^kB=0 \Rightarrow QA^kB=0.$



• $\widehat{C}(sI_{2n} - \widehat{A})^{-1}\widehat{B} \equiv 0 \Leftrightarrow \widehat{C}\widehat{A}^k\widehat{B} = 0$ for all $k \in \{0, 1, 2, \cdots\}$. • Note that

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Autonomy?

 $\widehat{C}\widehat{A}^k\widehat{B}=0 \Rightarrow (A^kB)^TQA^kB=0 \Rightarrow QA^kB=0.$

• Thus for $\widehat{C}(sI_{2n}-\widehat{A})^{-1}\widehat{B}\equiv 0$ we must have

$$Q\begin{bmatrix} B & AB & \cdots & A^{n-1}B\end{bmatrix} = 0.$$



$$(sI_{2n} - \widehat{A})^{-1}\widehat{B} \equiv 0 \Leftrightarrow \widehat{C}\widehat{A}^k\widehat{B} = 0$$
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 $\widehat{C}\widehat{A}^k\widehat{B}=0 \Rightarrow (A^kB)^TQA^kB=0 \Rightarrow QA^kB=0.$

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COMPUTATIONAL METHODS IN SYSTEMS AND CONTROL THEORY

$$Q\begin{bmatrix} B & AB & \cdots & A^{\mathbf{n}-1}B\end{bmatrix} = 0.$$

Since (A, B) is controllable, this is not possible unless Q = 0.

}.



Theorem (Main result)

Corresponding to a singular LQR problem with R = 0 (zero input cost) the following statements are equivalent:

- 1. $\widehat{C}(sI_{2n}-\widehat{A})^{-1}\widehat{B}\not\equiv 0.$
- 2. CGCARE is not solvable.



Theorem (Main result)

Corresponding to a singular LQR problem with R = 0 (zero input cost) the following statements are equivalent:

- 2. CGCARE is not solvable.
- 3. There exists no proportional state-feedback controller that solves the singular LQR Problem.



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Corresponding to a singular LQR problem with R = 0 (zero input cost) the following statements are equivalent:

- 1. $\widehat{C}(sI_{2n}-\widehat{A})^{-1}\widehat{B}\not\equiv 0.$
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$$\frac{d}{dt}x = \begin{bmatrix} 1 & 0 & 1\\ 1 & 0 & 1\\ 1 & 1 & 0 \end{bmatrix} x + \begin{bmatrix} 0\\ 1\\ 0 \end{bmatrix} u$$

Cost:
$$\int_0^\infty (x^T Q x) dt,$$
$$Q := \begin{bmatrix} 0 & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 1 \end{bmatrix}.$$

Example

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COMPUTATIONAL METHODS IN SYSTEMS AND CONTROL THEORY Example

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• CGCARE: $A^TK + KA + Q = 0$ and $\ker(R) \subseteq \ker(KB) \Rightarrow KB = 0$.



Example



COMPUTATIONAL METHODS IN Example

$$\begin{array}{l} \frac{d}{dt}x = \begin{bmatrix} 1 & 0 & 1\\ 1 & 0 & 1\\ 1 & 1 & 0 \end{bmatrix} x + \begin{bmatrix} 0\\ 1\\ 0 \end{bmatrix} u \\ \begin{array}{l} \text{Cost: } \int_{0}^{\infty} \left(x^{T}Qx\right) dt, \\ Q := \begin{bmatrix} 0 & 0 & 0\\ 0 & 0 & 0\\ 0 & 0 & 1 \end{bmatrix}. \\ \\ \text{CGCARE: } A^{T}K + KA + Q = 0 \text{ and } \ker(R) \subseteq \ker(KB) \Rightarrow KB = 0. \\ \\ \text{Let } K = \begin{bmatrix} k_{1} & k_{2} & k_{4}\\ k_{2} & k_{3} & k_{5}\\ k_{4} & k_{5} & k_{6} \end{bmatrix}. \end{array}$$

From the constrained equation KB = 0: $k_2 = k_3 = k_5 = 0$.



COMPUTATIONAL METHODS IN SYSTEMS AND CONTROL THEORY Example

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From the constrained equation KB = 0: $k_2 = k_3 = k_5 = 0$. We have from $A^TK + KA + Q = 0$:

$$\begin{bmatrix} 2(k_1+k_4) & k_4 & k_1+k_4+k_6\\ k_4 & 0 & k_6\\ k_1+k_4+k_6 & k_6 & 2k_4+1 \end{bmatrix} = 0$$



COMPUTATIONAL METHODS IN Example

$$\frac{d}{dt}x = \begin{bmatrix} 1 & 0 & 1\\ 1 & 0 & 1\\ 1 & 1 & 0 \end{bmatrix} x + \begin{bmatrix} 0\\ 1\\ 0 \end{bmatrix} u$$

$$Cost: \int_{0}^{\infty} (x^{T}Qx) dt,$$

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$$CGCARE: A^{T}K + KA + Q = 0 \text{ and } \ker(R) \subseteq \ker(KB) \Rightarrow KB = 0.$$

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CGCARE not solvable.

No P state-feedback controller



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PD controller

$$\frac{d}{dt}x = \begin{bmatrix} 1 & 0 & 1\\ 1 & 0 & 1\\ 1 & 1 & 0 \end{bmatrix} x + \begin{bmatrix} 0\\ 1\\ 0 \end{bmatrix} u$$

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Cost:
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COMPUTATIONAL METHODS IN SYSTEMS AND CONTROL THEORY PD controller

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 Cost: $\int_0^\infty (x^T Q x) dt$,
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• Eigenvalues of (E, H) are $\{-1, +1\}$.

$$E = \begin{bmatrix} I_{\mathbf{n}} & 0 & 0\\ 0 & I_{\mathbf{n}} & 0\\ 0 & 0 & 0 \end{bmatrix}, \ H = \begin{bmatrix} A & 0 & B\\ -Q & -A^T & 0\\ 0 & B^T & 0 \end{bmatrix}$$



 $V := \begin{bmatrix} 1 & 1 & -2 & 2 & 0 & 0 \end{bmatrix}$

such that $HV = EV\Gamma$, where $\Gamma = -1$. Define $V_1 = \begin{bmatrix} 1 & 1 & -2 \end{bmatrix}^T$, $V_3 = 0$.



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PD controller

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COMPUTATIONAL METHODS IN SYSTEMS AND CONTROL THEORY PD controller

$$\begin{array}{l} \bullet \ \frac{d}{dt}x = \begin{bmatrix} 1 & 0 & 1\\ 1 & 0 & 1\\ 1 & 1 & 0 \end{bmatrix} x + \begin{bmatrix} 0\\ 1\\ 0 \end{bmatrix} u \qquad \qquad \begin{array}{l} \operatorname{Cost:} \ \int_{0}^{\infty} \left(x^{T}Qx \right) dt, \\ Q := \begin{bmatrix} 0 & 0 & 0\\ 0 & 0 & 0\\ 0 & 0 & 1 \end{bmatrix}. \\ \bullet \ V_{1} = \begin{bmatrix} 1 & 1 & -2 \end{bmatrix}^{T}, \ V_{3} = 0. \\ \bullet \ Define \ X_{1} = \begin{bmatrix} V_{1} & B & AB \end{bmatrix} = \begin{bmatrix} 1 & 0 & 0\\ 1 & 1 & 0\\ -2 & 0 & 1 \end{bmatrix}.$$

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• Define
$$X_1 = \begin{bmatrix} V_1 & B & AB \end{bmatrix} = \begin{bmatrix} 1 & 0 & 0 \\ 1 & 1 & 0 \\ -2 & 0 & 1 \end{bmatrix}$$

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Design the controllers:

 $F_{p} := \begin{bmatrix} V_{3} & f_{0} & f_{1} \end{bmatrix} X_{1}^{-1}$ and $F_{d} := \begin{bmatrix} 0 & 1 & -f_{0} \end{bmatrix} X_{1}^{-1}$. where $f_0, f_1 \in \mathbb{R}$.



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$$\frac{d}{dt}x = \begin{bmatrix} 1 & 0 & 1 \\ 1 & 0 & 1 \\ 1 & 1 & 0 \end{bmatrix} x + \begin{bmatrix} 0 \\ 1 \\ 0 \end{bmatrix} u$$

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where $f_0, f_1 \in \mathbb{R}$.
Chosing $f_0 = 0$ and $f_1 = f$: $F_{\mathbf{p}} = \begin{bmatrix} 2f & 0 & f \end{bmatrix}$ and $F_{\mathbf{d}} = \begin{bmatrix} -1 & 1 & 0 \end{bmatrix}.$

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•
$$\frac{d}{dt}x = \begin{bmatrix} 1 & 0 & 1 \\ 1 & 0 & 1 \\ 1 & 1 & 0 \end{bmatrix} x + \begin{bmatrix} 0 \\ 1 \\ 0 \end{bmatrix} u$$

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Cost:
$$\int_0^\infty \left(x^T Q x \right) dt,$$
$$Q := \begin{bmatrix} 0 & 0 & 0 \\ 0 & 0 & 0 \\ 0 & 0 & 1 \end{bmatrix}.$$

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• Define
$$X_1 = \begin{bmatrix} V_1 & B & AB \end{bmatrix} = \begin{bmatrix} 1 & 0 & 0 \\ 1 & 1 & 0 \\ -2 & 0 & 1 \end{bmatrix}$$

- -

Design the controllers:

-

$$F_{p} := \begin{bmatrix} V_{3} & f_{0} & f_{1} \end{bmatrix} X_{1}^{-1} \text{ and } F_{d} := \begin{bmatrix} 0 & 1 & -f_{0} \end{bmatrix} X_{1}^{-1}.$$

-

where $f_0, f_1 \in \mathbb{R}$.

Chosing
$$f_0 = 0$$
 and $f_1 = f$: $F_p = \begin{bmatrix} 2f & 0 & f \end{bmatrix}$ and $F_d = \begin{bmatrix} -1 & 1 & 0 \end{bmatrix}$.

PD controller: $u = F_p x + F_d \dot{x}$.



•
$$\frac{d}{dt}x = \begin{bmatrix} 1 & 0 & 1\\ 1 & 0 & 1\\ 1 & 1 & 0 \end{bmatrix} x + \begin{bmatrix} 0\\ 1\\ 0 \end{bmatrix} u$$

• $V_1 = \begin{bmatrix} 1 & 1 & -2 \end{bmatrix}^T$, $V_3 = 0$.

_

Cost:
$$\int_0^\infty (x^T Q x) dt$$
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- PD controller: $u = F_p x + F_d \dot{x}$.
- Closed loop system: $(I_n BF_d)\frac{d}{dt}x = (A + BF_p)x$.



•
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• $V_1 = \begin{bmatrix} 1 & 1 & -2 \end{bmatrix}^T, V_3 = 0.$

Cost:
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Design the controllers:

-

$$F_{p} := \begin{bmatrix} V_{3} & f_{0} & f_{1} \end{bmatrix} X_{1}^{-1}$$
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where $f_0, f_1 \in \mathbb{R}$.

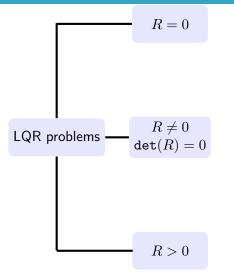
Chosing
$$f_0 = 0$$
 and $f_1 = f$: $F_p = \begin{bmatrix} 2f & 0 & f \end{bmatrix}$ and $F_d = \begin{bmatrix} -1 & 1 & 0 \end{bmatrix}$.

PD controller: $u = F_p x + F_d \dot{x}$.

Closed loop system: $(I_n - BF_d)\frac{d}{dt}x = (A + BF_p)x$. Choose $f: \det(s(I_n - BF_d) - (A + BF_p)) \neq 0$.

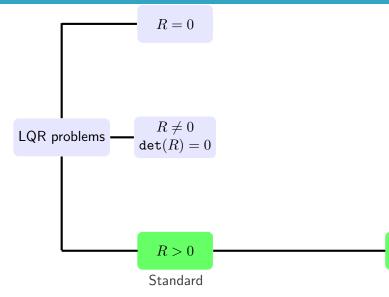


Conclusion





Conclusion



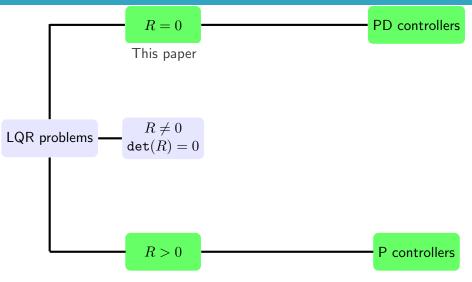
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P controllers

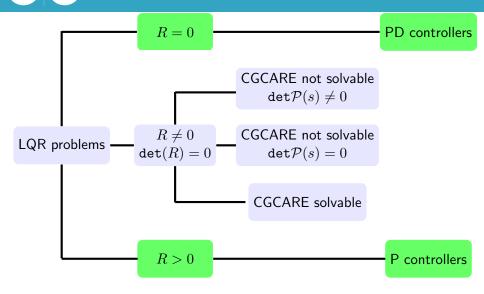


Conclusion



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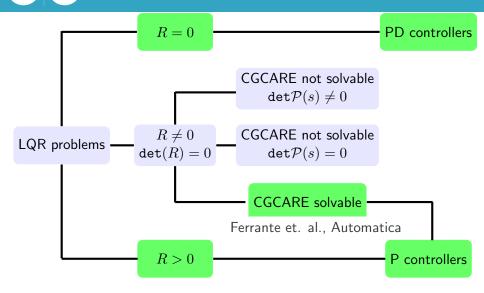
ODS IN Conclusion



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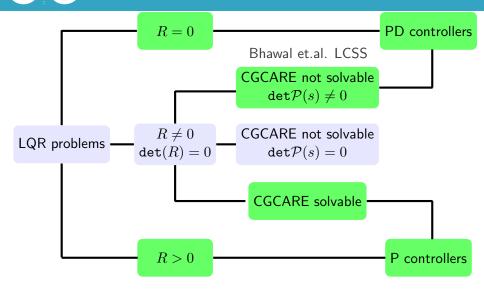
HODS IN Conclusion



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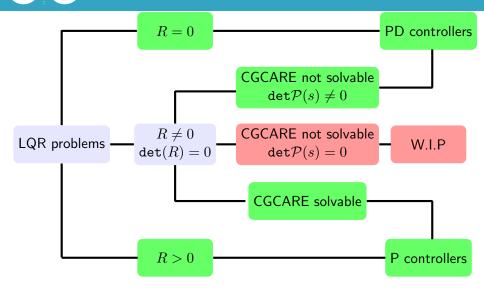
ONAL METHODS IN Conclusion



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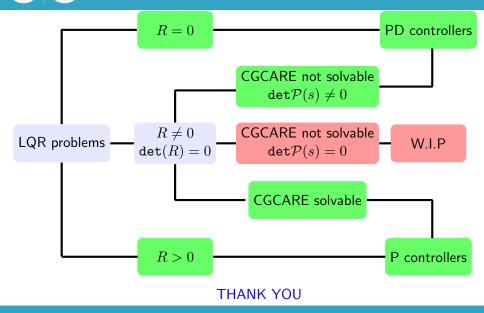
COMPUTATIONAL METHODS IN SYSTEMS AND CONTROL THEORY CONCLUSION



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